



MAX PLANCK INSTITUTE FOR DYNAMICS OF COMPLEX TECHNICAL SYSTEMS





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HRZZ Project Control of Dynamical Systems (ConDys)"

second project meeting -

Zagreb, 2-3 November 2017





- 1. Introduction
- 2. PMOR Methods based on Moment Matching
- 3. Optimal PMOR using Rational Interpolation?
- 4. Conclusions and Outlook



Parametric Dynamical Systems The Parametric Model Order Reduction (PMOR) Problem Error Measures

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Parametric Dynamical Systems

$$\Sigma(p): \begin{cases} E(p)\dot{x}(t;p) &= f(t,x(t;p),u(t),p), & x(t_0) = x_0, \\ y(t;p) &= g(t,x(t;p),u(t),p) \end{cases}$$
 (a)

with

- (generalized) states $x(t; p) \in \mathbb{R}^n$ ($E \in \mathbb{R}^{n \times n}$),
- inputs (controls) $u(t) \in \mathbb{R}^m$,
- outputs (measurements, quantity of interest) $y(t; p) \in \mathbb{R}^q$, (b) is called output equation,
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- $p \in \Omega \subset \mathbb{R}^d$ is a parameter vector, Ω is bounded.
- E(p) singular \Rightarrow (a) is system of differential-algebraic equations (DAEs) otherwise \Rightarrow (a) is system of ordinary differential equations (ODEs)





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Applications:

- Repeated simulation for varying material or geometry parameters, boundary conditions.
- control, optimization and design,
- of models, often generated by FE software (e.g., ANSYS, NASTRAN,...) or automatic tools (e.g., Modelica).





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Underlying PDE and boundary conditions often not accessible!

Parametric discretized model often not available,
but matrices for certain parameter values can be extracted

(or output data for given u and p can be generated!)



Linear, Time-Invariant (Parametric) Systems

$$E(p)\dot{x}(t;p) = A(p)x(t;p) + B(p)u(t), \quad A(p), \ E(p) \in \mathbb{R}^{n \times n},$$

$$y(t;p) = C(p)x(t;p), \qquad B(p) \in \mathbb{R}^{n \times m}, \ C(p) \in \mathbb{R}^{q \times n}.$$





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Laplace Transformation / Frequency Domain

Application of Laplace transformation

$$x(t; p) \mapsto x(s; p), \quad \dot{x}(t; p) \mapsto sx(s; p)$$

to linear system with $x(0; p) \equiv 0$:

$$sE(p)x(s;p) = A(p)x(s;p) + B(p)u(s), \quad y(s;p) = C(p)x(s;p),$$

yields I/O-relation in frequency domain:

$$y(s;p) = \left(\underbrace{C(p)(sE(p) - A(p))^{-1}B(p)}_{=:G(s,p)}\right)u(s).$$

G(s,p) is the parameter-dependent transfer function of $\Sigma(p)$.





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Goal: Fast evaluation of mapping $(u, p) \rightarrow y(s; p)$.





Problem

Approximate the dynamical system

$$\begin{array}{lcl} E(p)\dot{x} & = & A(p)x + B(p)u, & E(p), \ A(p) \in \mathbb{R}^{n \times n}, \\ y & = & C(p)x, & B(p) \in \mathbb{R}^{n \times m}, \ C(p) \in \mathbb{R}^{q \times n}, \end{array}$$

by reduced-order system

$$\begin{array}{cccc} \hat{E}(p)\dot{\hat{x}} & = & \hat{A}(p)\hat{x} + \hat{B}(p)u, & & \hat{E}(p), \ \hat{A}(p) \in \mathbb{R}^{r \times r}, \\ \hat{y} & = & \hat{C}(p)\hat{x}, & & \hat{B}(p) \in \mathbb{R}^{r \times m}, \ \hat{C}(p) \in \mathbb{R}^{q \times r}, \end{array}$$

of order $r \ll n$, such that

$$\|y - \hat{y}\| = \|Gu - \hat{G}u\| \le \|G - \hat{G}\| \cdot \|u\| < \text{tolerance} \cdot \|u\| \quad \forall \ p \in \Omega.$$





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 \implies Approximation problem: $\min_{\text{order } (\hat{G}) < r} \|G - \hat{G}\|.$





Structure Preservation

Parametric System

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Parametric model reduction goal:

preserve parameters as *symbolic quantities* in reduced-order model:

$$\widehat{\Sigma}(p): \left\{ \begin{array}{rcl} \widehat{E}(p)\dot{\widehat{x}}(t;p) & = & \widehat{A}(p)\widehat{x}(t;p) + \widehat{B}(p)u(t), \\ \widehat{y}(t;p) & = & \widehat{C}(p)\widehat{x}(t;p) \end{array} \right.$$

with states $\hat{x}(t; p) \in \mathbb{R}^r$ and $r \ll n$.





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Assuming parameter-affine representation:

$$E(p) = E_0 + e_1(p)E_1 + \dots + e_{q_E}(p)E_{q_E},$$

$$A(p) = A_0 + a_1(p)A_1 + \dots + a_{q_A}(p)A_{q_A},$$

$$B(p) = B_0 + b_1(p)B_1 + \dots + b_{q_B}(p)B_{q_B},$$

$$C(p) = C_0 + c_1(p)C_1 + \dots + c_{q_C}(p)C_{q_C},$$

allows easy parameter preservation for projection based model reduction.

Structure Preservation

Petrov-Galerkin-type projection

For given projection matrices $V, W \in \mathbb{R}^{n \times r}$ with $W^T V = I_r$ $(\rightsquigarrow (VW^T)^2 = VW^T$ is projector), compute

$$\hat{E}(p) = W^T E_0 V + e_1(p) W^T E_1 V + ... + e_{q_E}(p) W^T E_{q_E} V$$

$$\hat{A}(p) = W^T A_0 V + a_1(p) W^T A_1 V + \dots + a_{q_A}(p) W^T A_{q_A} V$$

$$\hat{B}(p) = W^T B_0 + b_1(p) W^T B_1 + \ldots + b_{q_B}(p) W^T B_{q_B}$$

$$\hat{C}(p) = C_0 V + c_1(p)C_1 V + \ldots + c_{q_c}(p)C_{q_c} V$$





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= \hat{B}_{0} + b_{1}(p) \hat{B}_{1} + \dots + b_{q_{B}}(p) \hat{B}_{q_{B}}
\hat{C}(p) = C_{0} V + c_{1}(p) C_{1} V + \dots + c_{q_{C}}(p) C_{q_{C}} V
= \hat{C}_{0} + c_{1}(p) \hat{C}_{1} + \dots + c_{q_{C}}(p) \hat{C}_{q_{C}}$$





Basis Generation — Global vs. Local

Local Bases

Obtain $V_k, W_k \in \mathbb{R}^{n \times r_k}$ using any non-parametric linear MOR method for a number of full-order models $\Sigma(p^{(k)})$, $k=1,\ldots,\ell$. Then compute reduced-order model by





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no need for affine parametrization, requires only system matrices $A(p^{(k)}), B(p^{(k)}), \ldots$

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- 2. transfer function interpolation: spurious poles of the parametric transfer function.
- 3. matrix interpolation: different models obtained in different coordinate systems → Procrustes problem → potential loss of accuracy; efficiency in "online" phase suffers from evaluating the interpolation operator.





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$$V := [V_1, \ldots, V_\ell], \qquad W := [W_1, \ldots, W_\ell]$$

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Avoids most of the problems encountered with local bases, but requires parameter-affine representation of system.



Empirical Matrix Interpolation Method

[B./Gugercin/Willcox 2015]

Given $V, W \in \mathbb{R}^{n \times r}$ and suppose only that $M(p) \in \mathbb{R}^{n \times t}$ can be evaluated at specific parameter values.



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- Goal: approximate $m(p) \approx \tilde{m}(p) = \Psi \alpha(p)$, where $\Psi \in \mathbb{R}^{nt \times \ell}$ and $\alpha(p) \in \mathbb{R}^{\ell}$ with $\ell \ll n$.



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- Then $\hat{m}(p) = \text{vec}(\hat{M}(p)) \in \mathbb{R}^{rt}$ (or \mathbb{R}^{r^2} if t = n) can be computed cheaply and independent of n as

$$\hat{m}(p) = \text{vec}\left(W^{T}M(p)V\right)$$

$$= (V^{T} \otimes W^{T})m(p) \approx (V^{T} \otimes W^{T})\tilde{m}(p) = (V^{T} \otimes W^{T})\Psi\alpha(p) = \tilde{m}(p).$$



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• This is achieved by sampling M(p) at $p = p^{(j)}$, $j = 1, \dots, \ell$, yielding

$$\psi_i = \text{vec}(M(p^{(i)}))$$
 and $\Psi = [\psi_1, \dots, \psi_\ell].$

Then apply (Q,D)EIM (or alike) to determine $\alpha(p)$ s.t. selected entries of $\tilde{m}(p)$ interpolate those entries of m(p).



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Let z_1, z_2, \ldots, z_ℓ be the selected indices to be exactly matched, and $Z := [e_{z_1}, \ldots, e_{z_\ell}]$. Then, forcing interpolation at the selected rows implies

$$Z^{T}m(p) = Z^{T}\Psi\alpha(p) \implies \alpha(p) = (Z^{T}\Psi)^{-1}Z^{T}m(p).$$





Empirical Matrix Interpolation Method

[B./Gugercin/Willcox 2015]

Given $V, W \in \mathbb{R}^{n \times r}$ and suppose only that $M(p) \in \mathbb{R}^{n \times t}$ can be evaluated at specific parameter values.

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- Undoing the vectorization yields the reduced model matrix

$$\hat{M}(p) := \operatorname{vec}^{-1}\left(\tilde{\hat{m}}(p)\right) = \operatorname{vec}^{-1}\left((V^T \otimes W^T)\Psi\alpha(p)\right) = \sum_{j=1}^{\ell} \alpha_j(p) \underbrace{W^T M(p^{(j)}) V}_{\text{precomputable}}$$



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Parametric Systems Norms

Mean-square error norm:

$$\|G-\hat{G}\|_{\mathcal{H}_2\otimes L_2(\Omega)}^2 := \frac{1}{2\pi} \int_{-\infty}^{+\infty} \int_{\mathbb{T}} \|G(\jmath\omega,\,p) - \hat{G}(\jmath\omega,\,p)\|_F^2 \,\mathrm{d}p_1 \ldots \mathrm{d}p_d \,\mathrm{d}\omega,$$

where $\|.\|_F$ denotes the Frobenius norm.

Worst-case error norm:

$$\|G - \hat{G}\|_{\mathcal{H}_{\infty} \otimes L_{\infty}(\Omega)} := \sup_{\omega \in \mathbb{R}} \|G(\jmath\omega, p) - \hat{G}(\jmath\omega, p)\|_{2}.$$



- 1. Introduction
- PMOR Methods based on Moment Matching Interpolatory Model Reduction PMOR based on Multi-Moment Matching
- 3. Optimal PMOR using Rational Interpolation?
- 4. Conclusions and Outlook



Computation of reduced-order model by projection

Given a linear (descriptor) system $E\dot{x} = Ax + Bu$, y = Cx with transfer function $G(s) = C(sE - A)^{-1}B$, a reduced-order model is obtained using truncation matrices $V, W \in \mathbb{R}^{n \times r}$ with $W^T V = I_r \ (\rightsquigarrow (VW^T)^2 = VW^T$ is projector) by computing

$$\hat{E} = W^T E V, \ \hat{A} = W^T A V, \ \hat{B} = W^T B, \ \hat{C} = C V.$$

Petrov-Galerkin-type (two-sided) projection: $W \neq V$,

Galerkin-type (one-sided) projection: W = V.



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Galerkin-type (one-sided) projection: W = V.

Rational Interpolation/Moment-Matching

Choose V, W such that

$$G(s_i) = \hat{G}(s_i), \quad j = 1, \ldots, k,$$

and

$$\frac{d^i}{ds^i}G(s_j)=\frac{d^i}{ds^i}\hat{G}(s_j), \quad i=1,\ldots,K_j, \quad j=1,\ldots,k.$$



Theorem (simplified) [GRIMME 1997, VILLEMAGNE/SKELTON 1987]

lf

$$\operatorname{span}\left\{ (s_1E - A)^{-1}B, \dots, (s_kE - A)^{-1}B \right\} \subset \operatorname{range}(V),$$

$$\operatorname{span}\left\{ (s_1E - A)^{-T}C^T, \dots, (s_kE - A)^{-T}C^T \right\} \subset \operatorname{range}(W),$$

then

$$G(s_j) = \hat{G}(s_j), \quad \frac{d}{ds}G(s_j) = \frac{d}{ds}\hat{G}(s_j), \quad \text{for } j = 1, \dots, k.$$





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Remarks:

computation of V, W from rational Krylov subspaces, e.g.,

- dual rational Arnoldi/Lanczos [GRIMME 1997],
- Iter. Rational Krylov-Alg. (IRKA) [Antoulas/Beattie/Gugercin 2006/08].



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Remarks:

using Galerkin/one-sided projection $(W \equiv V)$ yields $G(s_j) = \hat{G}(s_j)$, but in general

$$\frac{d}{ds}G(s_j)\neq \frac{d}{ds}\hat{G}(s_j).$$



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Remarks:

k = 1, standard Krylov subspace(s) of dimension K:

range
$$(V) = \mathcal{K}_K((s_1E - A)^{-1}, (s_1E - A)^{-1}B).$$

→ moment-matching methods/Padé approximation,

$$\frac{d^i}{ds^i}G(s_1)=\frac{d^i}{ds^i}\hat{G}(s_1), \quad i=0,\ldots,K-1(+K).$$



Comparison of Moment Matching and RBM

Numerical Example: A Printed Circuit Board (PCB)

System in time domain:

$$E\dot{x}(t) = Ax(t) + Bu(t),$$

 $y(t) = Cx(t).$

• System in frequency domain:

$$sEx(s) = Ax(s) + Bu(s),$$

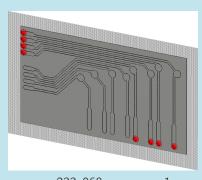
 $y(s) = Cx(s).$

Reduced basis method considers s
 as a parameter, and uses the system
 in frequency domain to compute

$$range(V) = span\{x(s_1), \dots, x(s_m)\}.$$

The ROM is obtained by Galerkin projection with V.

Printed circuit board



$$n = 233,060, m = q = 1.$$

Courtesy of TEMF, TU Darmstadt.

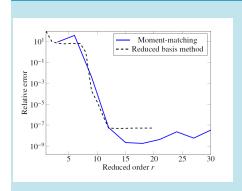


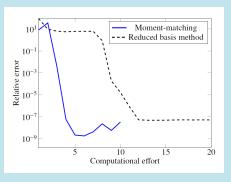


Comparison of Moment Matching and RBM

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Moment-matching vs. reduced basis method







Idea: choose appropriate frequency parameter \hat{s} and parameter vector \hat{p} , expand into multivariate power series about (\hat{s}, \hat{p}) and compute reduced-order model, so that

$$G(s,p) = \hat{G}(s,p) + \mathcal{O}\left(|s-\hat{s}|^K + \|p-\hat{p}\|^L + |s-\hat{s}|^k \|p-\hat{p}\|^I\right),$$

i.e., first K, L, k+l (mostly: K=L=k+l) coefficients (multi-moments) of Taylor/Laurent series coincide.



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Algorithms:

- [1] [Daniel et al. 2004]: explicit computation of moments, numerically unstable.
- [2] [FARLE ET AL. 2006/07]: Krylov subspace approach, only polynomial param.-dependance, numerical properties not clear, but appears to be robust.
- [3] [Weile et al. 1999, Feng/B. 2007/14]: Arnoldi-MGS method, employ recursive dependance of multi-moments, numerically robust, *r* often larger as for [2].
- [4] **New:** employ dual-weighted residual error bound and greedy procedure to define interpolation points an # of multi-moments matched

[Antoulas/B./Feng 2014/17].





Parametric System

Again, consider linear parametric system

$$\Sigma(p): \begin{cases} E(p)\dot{x}(t;p) &= A(p)x(t;p) + B(p)u(t), \\ y(t;p) &= C(p)x(t;p) \end{cases}$$

together with its transfer function G(s, p).





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For simplicity, assume $B(\mu) \equiv B$, and re-parameterize — $\mu := [s, p^T, \dots]^T \in \mathbb{C}^\ell$ such that with

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 $A(\mu) := sE(p) - A(p),$

we obtain linear-affine structure of $\mathcal{A}(\mu)$:

$$\mathcal{A}(\mu) = \mathcal{A}_0 + \mu_1 \mathcal{A}_1 + \ldots + \mu_\ell \mathcal{A}_\ell.$$



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In frequency domain, we may then re-write the parametric system as

$$A(\mu)x(\mu) = Bu(s), \qquad y(\mu) = C(\mu)x(\mu).$$



Multivariate Power Series Expansion I

Choose an expansion point $\mu^{(0)}$, and write

$$\mathcal{A}(\mu) = \mathcal{A}_0 + \mu_1 \mathcal{A}_1 + \dots + \mu_{\ell} \mathcal{A}_m$$

$$= \underbrace{(\mathcal{A}_0 + \mu_1^{(0)} \mathcal{A}_1 + \dots + \mu_{\ell}^{(0)} \mathcal{A}_m)}_{:=\mathcal{M}_0} + \Big((\mu_1 - \mu_1^{(0)}) \mathcal{A}_1 + \dots + (\mu_{\ell} - \mu_{\ell}^{(0)}) \mathcal{A}_{\ell} \Big)$$



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Using the Neumann lemma $((I-F)^{-1} = \sum_{j=0}^{\infty} F^j \text{ if } \|F\| < 1)$, we obtain

$$\begin{split} \mathcal{A}(\mu)^{-1} &= \sum_{j=0}^{\infty} (-1)^{j} \left((\mu_{1} - \mu_{1}^{(0)}) \mathcal{M}_{0}^{-1} \mathcal{A}_{1} + \ldots + (\mu_{\ell} - \mu_{\ell}^{(0)}) \mathcal{M}_{0}^{-1} \mathcal{A}_{\ell} \right)^{j} \mathcal{M}_{0}^{-1} \\ &= \sum_{j=0}^{\infty} (\sigma_{1} \mathcal{M}_{1} + \ldots \sigma_{\ell} \mathcal{M}_{\ell})^{j} \mathcal{M}_{0}^{-1}, \end{split}$$

where $\sigma_i = \mu_i - \mu_i^{(0)}$ and $\mathcal{M}_i = -\mathcal{M}_0^{-1} \mathcal{A}_i$ for $i = 1, \dots, \ell$.



Multivariate Power Series Expansion II

We have

$$A(\mu)x(\mu) = Bu(s).$$

and

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$$\begin{aligned} x(\mu) &= \mathcal{A}(\mu)^{-1} B u(s) = \sum_{j=0}^{\infty} (\sigma_1 \mathcal{M}_1 + \ldots + \sigma_{\ell} \mathcal{M}_{\ell})^j \underbrace{\mathcal{M}_0^{-1} B}_{=:\mathcal{B}} u(s) \\ &\approx \sum_{j=0}^{k} (\sigma_1 \mathcal{M}_1 + \ldots + \sigma_{\ell} \mathcal{M}_{\ell})^j \mathcal{B} u(s) =: \tilde{x}(\mu). \end{aligned}$$



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 $x(\mu)$ is (approximately, locally) contained in the Krylov subspace $\mathcal{K}_{k+1}((\sigma_1\mathcal{M}_1+\ldots\sigma_\ell\mathcal{M}_\ell),\mathcal{B})\Longrightarrow$

• Project the state-space onto this subspace.





- Project the state-space onto this subspace.
- \bullet Obtain an orthogonal basis using block-Arnoldi-MGS [B./Feng 2007/14], or TOAR [Bai/Su 2008].





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- Approximation is only valid locally (convergence radius of Neumann series!) \leadsto use several expansion points $\mu^{(0)}, \ldots, \mu^{(h)}$, and concatenate (and truncate) the local bases to obtain a global basis.

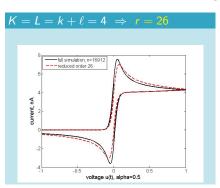


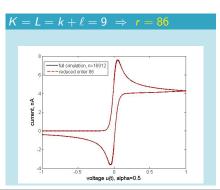
Numerical Examples: Electro-Chemical SEM

Compute cyclic voltammogram based on FE model

$$E\dot{x}(t) = (A_0 + p_1A_1 + p_2A_2)x(t) + Bu(t), \quad y(t) = c^Tx(t),$$

where $n = 16,912, m = 3, A_1, A_2$ diagonal.





Source: MOR Wiki: http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Scanning_Electrochemical_Microscopy

Open question

How to adaptively choose $\mu^{(i)}$?



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Possible approach: adopt ideas from Reduced Basis Methods, i.e., let

$$\|G(\mu) - \hat{G}(\mu)\| \le \Delta(\mu)$$
 or $\|y(\mu) - \hat{y}(\mu)\| \le \Delta_o(\mu)$

guide the selection of $\mu^{(i)}$ for computable *a posteriori* error bounds for the state or the output.



Error Bound for Automatic ROM Construction

Theorem (SISO case)

[Feng/Antoulas/B. 2015/17]

Assume that $\sigma_{\min}(\mathcal{G}(s,p))=:\beta(s,p)>0 \quad orall \operatorname{Re}(s)\geq 0, orall p\in \Omega$, then

$$|H(s,p) - \hat{H}(s,p)| \leq \tilde{\Delta}(s,p) + |(\hat{x}^{du})^H \mathbf{r}^{pr}(s,p)| =: \Delta(s,p),$$

where

$$\tilde{\Delta}(s,p) = \frac{\|r^{du}(s,p)\|_2 \|r^{pr}(s,p)\|_2}{\beta(s,p)},$$

with the primal and dual residuals r^{pr} , r^{du} and the reduced "dual state" $\hat{\chi}^{du}$:

$$r^{pr}(s,p) = \| (B - (sE(p) - A(p))) (V(s\hat{E}(p) - \hat{A}(p))^{-1}\hat{B}) \|,$$

$$r^{du}(s,p) = \| (C^{T} - (\bar{s}E(p) - A(p))^{T}) \hat{x}^{du} \|,$$

$$\hat{x}^{du} = -V^{du}(\bar{s}\hat{E}^{du}(p) - \hat{A}^{du}(p))^{-T}\hat{C}^{du}.$$

The dual reduced-order system is computed using Galerkin projection with V^{du} obtained by applying multi-moment matching algorithm to "dual" system $(\bar{s}E(p)^T - A(p)^T, C^T)$.



Error Bound for Automatic ROM Construction

Remarks

- For application in "RBM fashion", $r^{du}(\mu)$, $r^{pr}(\mu)$ can be efficiently computed, need to solve sparse linear systems on training set, i.e., one sparse factorization for each sampling point.
- $\beta(s,p) = \sigma_{\min}(G(s,p))$ easily computable on the training set system solves for evaluation of the transfer function readily available from residual computation!
- Extension to MIMO case possible taking max over all I/O channels.
- Can use Petrov-Galerkin framework using $W = V^{du}$ at no extra cost!

Algorithm 1 Automatic generation of the ROM: adaptively selecting $\mu^{(i)}$

Input: $V = [\]; \ \epsilon > \epsilon_{tol};$ Initial expansion point: $\hat{\mu}; \ i := -1;$ $\Xi_{\text{train}}:$ a set of samples of μ covering the parameter domain. **Output:** V.

- 1: while $\epsilon > \epsilon_{tol}$ do
- 2: i = i + 1;
- 3: $\mu^{(i)} = \hat{\mu};$
- 4: $V_{\mu^{(i)}} = \text{orthogonal basis of } \mathcal{K}_{k+1}((\sigma_1^{(i)}\mathcal{M}_1 + \ldots + \sigma_\ell^{(i)}\mathcal{M}_\ell), \mathcal{B});$
- 5: $V = \operatorname{orth}([V, V_{\mu^{(i)}}]);$
- 6: $\hat{\mu} = \arg \max_{\mu \in \Xi_{train}} \Delta(\mu);$
- 7: $\epsilon = \Delta(\hat{\mu});$
- 8: end while





Numerical Example: Silicon Nitride Membrane

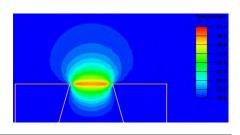
A SiN membrane can be a part of a gas sensor, an infra-red sensor, a microthruster, etc. Heat tansfer in the membrane is described by

$$(E_0 + \rho c_p E_1) \dot{x}(t) = -(K_0 + \kappa K_1 + h K_2) x(t) + b u(t) y(t) = C x(t),$$

with parameters

- density $\rho \in [3000, 3200]$,
- specific heat capacity $c_p \in [400, 750]$,
- thermal conductivity $\kappa \in [2.5, 4]$,
- membrane heat transfer coefficient h ∈ [10, 12].

and frequency $f \in [0, 25]Hz$.



Source: MOR Wiki: http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Silicon_nitride_membrane



Numerical Examples: Silicon Nitride Membrane

Setting

- Training set: $\Xi_{train} = 5$ random samples for ρ and c_{ρ} , 3 random samples for κ and h, respectively, 10 samples of Laplace variable s.
- Error measures:

$$arepsilon_{ ext{true}}^{ ext{rel}} = \max_{\mu \in \Xi_{ ext{train}}} |G(\mu) - \hat{G}(\mu)| / |G(\mu)|,$$

$$\Delta^{rel}(\mu) = \Delta(\mu)/|\hat{G}(\mu)|$$

•
$$V_{\mu^{(i)}} = \operatorname{span}\{\mathcal{B}, (\sigma_1^{(i)}\mathcal{M}_1 + \ldots + \sigma_\ell^{(i)}\mathcal{M}_\ell)\mathcal{B}\}, \ \epsilon_{tol}^{re} = 10^{-2}, \ n = 60,020, \ r = 8.$$

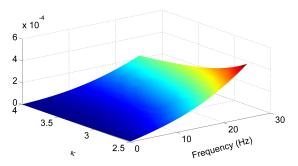
iter.	$arepsilon_{true}^{rel}$	$\Delta^{rel}(\mu^{(i)})$	S	ρc_p	κ	h
1	1×10^{-3}	3.44	18.94	1.37×10^{6}	2.74	10.97
2	1×10^{-4}	4.59×10^{-2}	0.89	1.31×10^6	3.96	11.60
3	2.80×10^{-5}	4.07×10^{-2}	23.98	2.35×10^6	3.94	10.28
4	2.58×10^{-6}	2.62×10^{-5}	0.89	2.31×10^6	2.74	10.28





Numerical Examples: Silicon Nitride Membrane

Verification of the accuracy of the ROM for κ over set Ξ_{fine} with 16 equidistant samples of κ , 51 equidistant samples of the frequency f, while the other parameters are fixed.

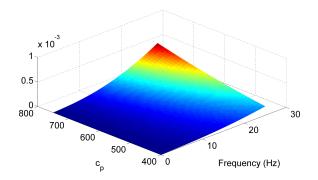


Relative error of the final ROM changing with κ and frequency.



Numerical Examples: Silicon Nitride Membrane

Verification of the accuracy of the ROM for c_p over set Ξ_{fine} with 36 equidistant samples of c_p , 51 equidistant samples of the frequency f, while the other parameters are fixed.

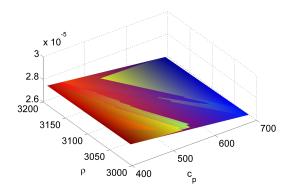


Relative error of the final ROM changing with c_p and frequency.



Numerical Examples: Silicon Nitride Membrane

Verification of the accuracy of the ROM for ρ , c_p over set Ξ_{fine} with 50 random samples of ρ , c_p , respectively, the other parameters are fixed.



Relative error of the final ROM changing with c_p and κ .



- 1. Introduction
- 2. PMOR Methods based on Moment Matching
- 3. Optimal PMOR using Rational Interpolation? \mathcal{H}_2 -optimal Model Reduction for Linear Systems \mathcal{H}_2 -(sub)optimal Model Reduction for Linear Parametric Systems \mathcal{H}_2 -optimal Model Reduction for Special Linear Parametric Systems A Comparison of PMOR Methods
- 4. Conclusions and Outlook



Greedy expansion point selection has a heuristic nature and relies on a training set.





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How to determine the right number of partial derivatives to be matched at the expansion points is an open problem (for potential solutions in the non-parametric case, see [Feng/Korvink/B. 2015, Bonin/Fassbender/Soppa/Zäh 2016, Lee/Chu/Feng 2006,...].





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Hence, we investigate the problem: for a given order r of the reduced-order model, can we provide necessary conditions for a rational interpolant to minimize

$$\|G - \hat{G}\|_{\mathcal{H}_2 \otimes L_2(\Omega)}^2 := \frac{1}{2\pi} \int_{-\infty}^{+\infty} \int\limits_{\Omega} \|G(\jmath\omega, p) - \hat{G}(\jmath\omega, p)\|_F^2 dp_1 \dots dp_d d\omega ?$$





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Following the non-parametric case, one would need:

ullet Projection-based framework for tangential rational interpolation. $[\checkmark]$



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Following the non-parametric case, one would need:

- Projection-based framework for tangential rational interpolation. [✓]
- Iterative procedure for selecting interpolation points. $[\chi] \dots [\sqrt{\ }]$ for special case.





H₂-Model Reduction for Linear Systems

Consider stable (i.e. $\Lambda(A) \subset \mathbb{C}^-$) linear systems Σ ,

$$\dot{x}(t) = Ax(t) + Bu(t), \ y(t) = Cx(t)$$
 $\simeq y(s) = \underbrace{C(sI - A)^{-1}B}_{=:G(s)} u(s)$

System norms

Recall: two common system norms for measuring approximation quality are

•
$$\mathcal{H}_2$$
-norm, $\|\Sigma\|_{\mathcal{H}_2} = \left(\frac{1}{2\pi} \int_0^{2\pi} \operatorname{tr}\left(\left(G^T(-\jmath\omega)G(\jmath\omega)\right)\right) d\omega\right)^{\frac{1}{2}}$,

$$\bullet \ \mathcal{H}_{\infty}\text{-norm, } \|\Sigma\|_{\mathcal{H}_{\infty}} = \sup_{\omega \in \mathbb{R}} \sigma_{\mathsf{max}} (\mathit{G}(\jmath \omega)),$$

where

$$G(s) = C(sI - A)^{-1} B.$$



Error system and \mathcal{H}_2 -Optimality

[Meier/Luenberger 1967]

In order to find an \mathcal{H}_2 -optimal reduced system, consider the error system $G(s) - \hat{G}(s)$ which can be realized by

$$A^{err} = \begin{bmatrix} A & 0 \\ 0 & \hat{A} \end{bmatrix}, \quad B^{err} = \begin{bmatrix} B \\ \hat{B} \end{bmatrix}, \quad C^{err} = \begin{bmatrix} C & -\hat{C} \end{bmatrix}.$$



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Assuming a coordinate system in which \hat{A} is diagonal and taking derivatives of

$$||G(.) - \hat{G}(.)||_{\mathcal{H}_2}^2$$

with respect to free parameters in $\Lambda(\hat{A}), \hat{B}, \hat{C} \leadsto$ first-order necessary \mathcal{H}_2 -optimality conditions (SISO)

$$G(-\hat{\lambda}_i) = \hat{G}(-\hat{\lambda}_i),$$

$$G'(-\hat{\lambda}_i) = \hat{G}'(-\hat{\lambda}_i),$$

where $\hat{\lambda}_i$ are the poles of the reduced system $\hat{\Sigma}$.



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First-order necessary \mathcal{H}_2 -optimality conditions (MIMO):

$$G(-\hat{\lambda}_i)\tilde{B}_i = \hat{G}(-\hat{\lambda}_i)\tilde{B}_i, \qquad \text{for } i = 1, \dots, r,$$

$$\tilde{C}_i^T G(-\hat{\lambda}_i) = \tilde{C}_i^T \hat{G}(-\hat{\lambda}_i), \qquad \text{for } i = 1, \dots, r,$$

$$\tilde{C}_i^T H'(-\hat{\lambda}_i)\tilde{B}_i = \tilde{C}_i^T \hat{G}'(-\hat{\lambda}_i)\tilde{B}_i \qquad \text{for } i = 1, \dots, r,$$

where $\hat{A} = R\hat{\Lambda}R^{-T}$ is the spectral decomposition of the reduced system and $\tilde{B} = \hat{B}^T R^{-T}$, $\tilde{C} = \hat{C}R$.



Error system and \mathcal{H}_2 -Optimality

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First-order necessary \mathcal{H}_2 -optimality conditions (MIMO):

$$G(-\hat{\lambda}_{i})\tilde{B}_{i} = \hat{G}(-\hat{\lambda}_{i})\tilde{B}_{i}, \qquad \text{for } i = 1, \dots, r,$$

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$$\tilde{C}_{i}^{T}H'(-\hat{\lambda}_{i})\tilde{B}_{i} = \tilde{C}_{i}^{T}\hat{G}'(-\hat{\lambda}_{i})\tilde{B}_{i} \qquad \text{for } i = 1, \dots, r,$$

$$\Leftrightarrow \text{vec}(I_{q})^{T}\left(e_{j}e_{i}^{T}\otimes C\right)\left(-\hat{\Lambda}\otimes I_{n} - I_{r}\otimes A\right)^{-1}\left(\tilde{B}^{T}\otimes B\right)\text{vec}(I_{m})$$

$$= \text{vec}(I_{q})^{T}\left(e_{j}e_{i}^{T}\otimes\hat{C}\right)\left(-\hat{\Lambda}\otimes I_{r} - I_{r}\otimes\hat{A}\right)^{-1}\left(\tilde{B}^{T}\otimes\hat{B}\right)\text{vec}(I_{m}),$$

$$\text{for } i = 1, \dots, r \text{ and } j = 1, \dots, q.$$



Interpolation of the Transfer Function [GRIMME 1997]

Construct reduced transfer function by Petrov-Galerkin projection $\mathcal{P} = VW^T$, i.e.

$$\hat{G}(s) = CV (sI - W^T AV)^{-1} W^T B,$$





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where V and W are given as

$$V = [(-\mu_1 I - A)^{-1} B, \dots, (-\mu_r I - A)^{-1} B],$$

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for i = 1, ..., r.

Starting with an initial guess for $\hat{\Lambda}$ and setting $\mu_i \equiv \hat{\lambda}_i \leadsto$ iterative algorithms (IRKA/MIRIAm) that yield \mathcal{H}_2 -optimal models.

[Gugercin et al. 2006/08], [Bunse-Gerstner et al. 2007], [Van Dooren et al. 2008]



The Basic IRKA Algorithm

Algorithm 2 IRKA (MIMO version/MIRIAm)

Input: A stable, B, C, \hat{A} stable, \hat{B} , \hat{C} , $\delta > 0$.

Output: A^{opt}, B^{opt}, C^{opt}

1: while
$$(\max_{j=1,...,r}\left\{\frac{|\mu_j-\mu_j^{\rm old}|}{|\mu_j|}\right\}>\delta)$$
 do

- 2: diag $(\mu_1, \dots, \mu_r) := R^{-1} \hat{A} R$ = spectral decomposition.
- 3: $\tilde{B} = \hat{B}^H R^{-T}$, $\tilde{C} = \hat{C}R$.

4:
$$V = [(-\mu_1 I - A)^{-1} B \tilde{b}_1, \dots, (-\mu_r I - A)^{-1} B \tilde{b}_r]$$

5:
$$W = [(-\mu_1 I - A^T)^{-1} C^T \tilde{c}_1, \dots, (-\mu_r I - A^T)^{-1} C^T \tilde{c}_r]$$

6:
$$V = \text{orth}(V), W = \text{orth}(W), W = W(V^H W)^{-1}$$

7:
$$\hat{A} = W^H A V$$
, $\hat{B} = W^H B$, $\hat{C} = C V$.

- 8: end while
- 9: $A^{opt} = \hat{A}$, $B^{opt} = \hat{B}$, $C^{opt} = \hat{C}$.





Theory: Interpolation of the Transfer Function

Theorem

[Baur/Beattie/B./Gugercin 2007/11]

Let

$$\hat{G}(s,p) := \hat{C}(p)(s\hat{E}(p) - \hat{A}(p))^{-1}\hat{B}(p)
= C(p)V(sW^{T}E(p)V - W^{T}A(p)V)^{-1}W^{T}B(p).$$

Suppose $\hat{p} = [\hat{p}_1, ..., \hat{p}_d]^T$ and $\hat{s} \in \mathbb{C}$ are chosen such that both $\hat{s} E(\hat{p}) - A(\hat{p})$ and $\hat{s} E(\hat{p}) - \hat{A}(\hat{p})$ are invertible.

$$(\hat{s} E(\hat{p}) - A(\hat{p}))^{-1} B(\hat{p}) \in \operatorname{range}(V)$$

or

$$\left(C(\hat{p})\left(\hat{s}\,E(\hat{p})-A(\hat{p})\right)^{-1}\right)^T\in\mathrm{range}\left(W\right),$$

then $G(\hat{s}, \hat{p}) = \hat{G}(\hat{s}, \hat{p})$.





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then $G(\hat{s}, \hat{p}) = \hat{G}(\hat{s}, \hat{p})$.

Extension to MIMO case using tangential interpolation: let $0 \neq b \in \mathbb{R}^m$, $0 \neq c \in \mathbb{R}^q$.

- a) If $(\hat{s} E(\hat{p}) A(\hat{p}))^{-1} B(\hat{p})b \in \text{range}(V)$, then $G(\hat{s}, \hat{p})b = \hat{G}(\hat{s}, \hat{p})b$.
- b) If $\left(c^T C(\hat{p}) \left(\hat{s} E(\hat{p}) A(\hat{p})\right)^{-1}\right)^T \in \text{range}(W)$, then $c^T G(\hat{s}, \hat{p}) = c^T \hat{G}(\hat{s}, \hat{p})$.





Theory: Interpolation of the Parameter Gradient

Theorem

[Baur/Beattie/B./Gugercin 2007/11]

Suppose that E(p), A(p), B(p), C(p) are C^1 in a neighborhood of $\hat{p} = [\hat{p}_1, \dots, \hat{p}_d]^T$ and that both $\hat{s} E(\hat{p}) - A(\hat{p})$ and $\hat{s} E(\hat{p}) - \hat{A}(\hat{p})$ are invertible.

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then

$$\nabla_{p}G(\hat{s},\hat{p})=\nabla_{p}G_{r}(\hat{s},\hat{p}), \qquad \frac{\partial}{\partial s}G(\hat{s},\hat{p})=\frac{\partial}{\partial s}\hat{G}(\hat{s},\hat{p}).$$

Note: result extends to MIMO case using tangential interpolation: Let $0 \neq b \in \mathbb{R}^m$, $0 \neq c \in \mathbb{R}^q$ be arbitrary. If $(\hat{s} E(\hat{p}) - A(\hat{p}))^{-1} B(\hat{p}) b \in \operatorname{range}(V)$ and $\left(c^T C(\hat{p}) (\hat{s} E(\hat{p}) - A(\hat{p}))^{-1}\right)^T \in \operatorname{range}(W)$, then $\nabla_p c^T G(\hat{s}, \hat{p}) b = \nabla_p c^T \hat{G}(\hat{s}, \hat{p}) b$.





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abla_p G_r(\hat{s}, \hat{p}), \qquad \frac{\partial}{\partial s} G(\hat{s}, \hat{p}) = \frac{\partial}{\partial s} \hat{G}(\hat{s}, \hat{p}).$$

- 1. Reduced-order model satisfies necessary conditions for surrogate models in trust region methods [Alexandrov/Dennis/Lewis/Torczon 1998].
- 2. Approximation of gradient allows use of reduced-order model for sensitivity analysis.





Generic implementation of interpolatory PMOR

Define A(s, p) := sE(p) - A(p).

- 1. Select "frequencies" $s_1, \ldots, s_k \in \mathbb{C}$ and parameter vectors $p^{(1)}, \ldots, p^{(\ell)} \in \Omega \subset \mathbb{R}^d$.
- 2. Compute (orthonormal) basis of

$$\mathcal{V} = \mathrm{span}\, \big\{ \mathcal{A}(s_1, p^{(1)})^{-1} B(p^{(1)}), \dots, \mathcal{A}(s_k, p^{(\ell)})^{-1} B(p^{(\ell)}) \big\}.$$

3. Compute (orthonormal) basis of

$$\mathcal{W} = \operatorname{span}\big\{\mathcal{A}(s_1, \boldsymbol{\rho}^{(1)})^{-T} \, \boldsymbol{C}(\boldsymbol{\rho}^{(1)})^T, \dots, \mathcal{A}(s_k, \boldsymbol{\rho}^{(\ell)})^{-T} \, \boldsymbol{C}(\boldsymbol{\rho}^{(\ell)})^T\big\}.$$

- 4. Set $V := [v_1, \dots, v_{k\ell}], \ \tilde{W} := [w_1, \dots, w_{k\ell}], \ \text{and} \ W := \tilde{W}(\tilde{W}^T V)^{-1}.$ (Note: $r = k\ell$).
- 5. Compute $\begin{cases} \hat{A}(p) := W^T A(p) V, & \hat{B}(p) := W^T B(p) V, \\ \hat{C}(p) := W^T C(p) V, & \hat{E}(p) := W^T E(p) V. \end{cases}$





Remarks

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• Optimal choice of interpolation frequencies s_k and parameter vectors $p^{(k)}$ possible for special cases.





Numerical Example: Thermal Conduction in a Semiconductor Chip

- Important requirement for a compact model of thermal conduction is boundary condition independence.
- The thermal problem is modeled by the heat equation, where heat exchange through device interfaces is modeled by convection boundary conditions containing film coefficients $\{p_i\}_{i=1}^3$ describing the heat exchange at *i*th interface.
- Spatial semi-discretization leads to

$$E\dot{x}(t) = (A_0 + \sum_{i=1}^{3} p_i A_i)x(t) + bu(t), \quad y(t) = c^{T}x(t),$$

where n = 4,257, A_i , i = 1,2,3, are diagonal.

Source: C.J.M Lasance, Two benchmarks to facilitate the study of compact thermal modeling phenomena, IEEE Transactions on Components and Packaging Technologies, 24(4):559–565, 2001.

MOR Wiki: http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Microthruster_Unit



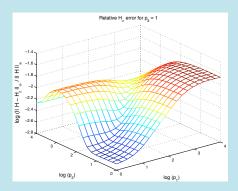


\mathcal{H}_2 -(sub)optimal Model Reduction for Linear Parametric Systems

Numerical Example: Thermal Conduction in a Semiconductor Chip

Choose 2 interpolation points for parameters ("important" configurations), 8/7 H_2 -optimal interpolation frequencies selected by IRKA. $\Rightarrow k=2, \ell=8,7$, hence r=15.

$$p_3 = 1$$
, $p_1, p_2 \in [1, 10^4]$.







\mathcal{H}_2 -optimal Model Reduction for Special Linear Parametric Systems

Optimality of Interpolation Points

Theorem

[Baur/Beattie/B./Gugercin 2011]

For special parameterized SISO systems,

$$A(p) \equiv A_0, \ E(p) \equiv E_0, \ B(p) = B_0 + p_1 B_1, \ C(p) = C_0 + p_2 C_1,$$

optimal choice possible, necessary conditions:

If \hat{G} minimizes the approximation error w.r.t. $\|G - \hat{G}\|_{\mathcal{H}_2 \times L_2(\Omega)}$, $p \in \Omega \subset \mathbb{R}^d$, and $\Lambda(\hat{A}, \hat{E}) = \{\hat{\lambda}_1, \dots, \hat{\lambda}_r\}$ (all simple), then the interpolation frequencies satisfy

$$s_i = -\hat{\lambda}_i, \quad i = 1, \ldots, r,$$

and the parameter interpolation points $\{p^{(1)},\dots,p^{(r)}\}$ satisfy the interpolation conditions

$$G(-\hat{\lambda}_{k}, p^{(k)}) = \hat{G}(-\hat{\lambda}, p^{(k)}),$$

$$\frac{\partial}{\partial s}G(-\hat{\lambda}, p^{(k)}) = \frac{\partial}{\partial s}\hat{G}(-\hat{\lambda}, p^{(k)}), \quad \nabla_{p}G(-\hat{\lambda}, p^{(k)}) = \nabla_{p}\hat{G}(-\hat{\lambda}, p^{(k)}).$$





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Proof:

$$\|G\|_{\mathcal{H}_2\times L_2(\Omega)} = \|L^T \tilde{G}L\|_{\mathcal{H}_2}, \quad \text{where } \tilde{G}(s) = \left[\begin{array}{c} C_0 \\ C_1 \end{array} \right] (sE-A)^{-1} [B_0,B_1], \ L = \left[\begin{array}{c} 1 & 0 \\ \frac{1}{2} & \frac{1}{2\sqrt{3}} \end{array} \right].$$

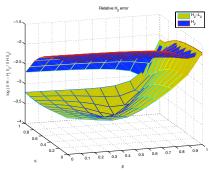
 \Longrightarrow Computation via IRKA applied to \tilde{G} .



\mathcal{H}_2 -optimal Model Reduction for Special Linear Parametric Systems

Optimality of Interpolation Points — Numerical Example

- Model for evolution of temperature distribution on a plate, described by the heat equation.
- FDM SISO model of order n = 197.
- Parameter $p_1 \in [0,1]$ encodes movement of heat source from B_0 to $B_0 + B_1$, analogous for relocation of measurement.



Relative $\mathcal{H}_2 \otimes L_2(\Omega)$ error: 7.5×10^{-4} .

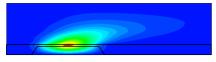




A Comparison of PMOR Methods: Anemometer

Consider an anemometer, a flow sensing device located on a membrane used in the context of minimizing heat dissipation.





Source: [BAUR/B./GREINER/KORVINK/LIENEMANN/MOOSMANN 2011]

• FE model:

$$E\dot{x}(t) = (A + pA_1)x(t) + Bu(t), \quad y(t) = Cx(t), \quad x(0) = 0,$$

• $n = 29,008, m = 1, q = 3, p_1 \in [0,1]$ fluid velocity.

Source: MOR Wiki: http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Anemometer



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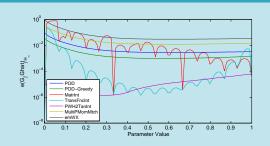
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\mathcal{H}_{∞} error





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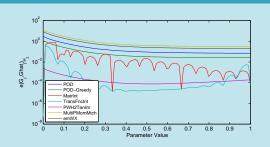
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\mathcal{H}_2 error



For more details of this comparisons, and other tests, see

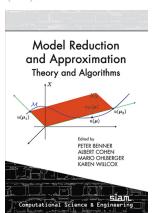


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Chapter 9 in





- 1. Introduction
- 2. PMOR Methods based on Moment Matching
- 3. Optimal PMOR using Rational Interpolation?
- 4. Conclusions and Outlook





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- Several extensions to nonlinear systems, but just starting.
- New direction: data-enhanced approaches, merging ideas from Loewner framework with model-based methods.
- Most of the methods can be used to significantly accelerate UQ by Monte Carlo or Stochastic Collocation methods!





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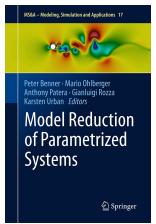
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